Experimental Study on MIMO Performance of Modulated Scattering Array Antenna in Indoor Environment

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Abstract: The modulated scattering array antenna (MSAA) is composed of one normal antenna element and several modulated scattering elements (MSE). In this report, a 2-element MSAA is used as the receiving antenna in a 2 by 2 multiple input multiple output (MIMO) system. The MIMO performance of the MSAA with various array spacing is measured to investigate experimentally the relation between the array spacing and the MIMO performance of the MSAA in the Line-of-Sight (LOS) and Non-Line-of-Sight (NLOS) indoor environment. It is found that the Error Vector Magnitude (EVM) and the Channel Capacity which reflect the MIMO performance can be affected by the array spacing. Furthermore, the EVM is complementary with the channel capacity.

Keywords: array antenna, modulation, mobile handsets, wireless communication, MIMO

1. INTRODUCTION

Multiple input multiple output (MIMO) communication system has become a promising technology for the nextgeneration wireless communications system, because it could achieve much higher spectral efficiency and transfer reliability than the conventional wireless communication techniques with the same transmitted power and frequency bandwidth [1]. However, it is very difficult to develop array antennas suitable for mobile handsets, because of some problems such as the limited space on the handset to mount array antennas with sufficiently low mutual coupling and correlation between antennas [2], [3]. Moreover, because a number of separate RF front-end circuits are required corresponding to the number of array elements, a large amount of packaging space for the RF front-end circuits is necessary. Therefore, it is essential to develop array antennas with simple configurations which are suitable for mobile handsets in MIMO communications.

A new concept of array antennas, which is called modulated scattering array antenna (MSAA), based on the modulated scattering technique (MST) has been proposed [4], [5]. The MSAA consists of one normal antenna element and several modulated scattering elements (MSE) without RF front-end circuit. The previous researches showed that the MSAA is suitable for mobile handsets in the MIMO communications where the space and the cost are limited because of its simple configuration [6].

It is apparent that reduction the array spacing between the normal antenna element and the MSE can increase the scattering signal, but high correlation due to the compact array spacing may degrade the MIMO performance. Therefore, we investigated further the MSAA in MIMO communications to see whether the MIMO performance of the MSAA for mobile handsets can be improved by regulating the array spacing. Experimental measurement was carried out to study the MIMO performance in the LOS and NLOS environment of an indoor 2 by 2 MIMO system where the MSAA was used as the receiving antenna. Because the Error Vector Magnitude (EVM) and the Channel Capacity reflect MIMO performance, they were measured and compared along with the different array spacing to study the relation between the array spacing and the MIMO performance of the MSAA.

The experimental configuration of the MIMO communication system is presented in Section 2. Section 3 shows the experimental results.

2. EXPERIMENTAL CONFIGURATION

The geometry of a 2-element dipole MSAA is shown in Fig. 1. The MSAA is composed of two half-wavelength dipole elements with array spacing of 0.1, 0.2, 0.3, 0.4, 0.5 and 0.6 wavelengths. In the MSAA, the left element is the passive dipole antenna and the right one is the MSE. A schottky diode is mounted at the center of the dipole which is used as the nonlinear impedance for modulation.





Fig. 1. Geometry of dipole antenna and modulated scattering dipole antenna

Fig. 2 shows the measurement system which was developed to demonstrate the MIMO performance of the MSAA in a 2 by 2 MIMO communication system operated with IEEE 802.11n protocol. Two log-periodic dipole array antennas with two wavelengths array spacing were used as the transmitting antennas. The Agilent 89600S vector signal analyzer with two 2.5 GHz RF input channels and software option 89601X-B7Z for IEEE 802.11n MIMO modulation analysis were used to receive the signals from the measured MSAA.



Fig. 2. 2-channal MIMO measurement system





The experiment was implemented in a 7.3×11.1 meter lecture room with concrete structure which is shown in Fig. 3. The distance between the transmitting and receiving antennas is about 10 meters in a LOS (line-of-sight) and NLOS (non-lineof-sight) environment. The location of transmitting antenna was fixed, while the receiving antenna was moved by a step of 5 cm in a 50 cm \times 50 cm area. Therefore, measurement was repeated 11 \times 11 times and the constellation diagrams of the demodulated IEEE 802.11n signals for 2 streams were recorded at each point. Further, the EVM is calculated for the constellation diagram for every location of the receiving antenna.

3. EXPERIMENTAL RESULTS

Fig. 4 shows the constellation diagram of 2 streams demodulated from IEEE 802.11n signals received by the MSAA, which includes QPSK-modulated data symbols and BPSK-modulated pilot symbols. It is shown that symbols of 2 streams are shifted slightly from their ideal location. The degradation of stream 2 is caused by the lower gain of the MSE as reported in [4] and [5], where it was found that the gain of the MSE element is usually 15-20 dB lower than that of the normal antenna element. Because the measurement was repeated 121 times while slightly changing the location of the receiving antenna, 121 values of the EVM were obtained and they were further expressed in the form of CDF. EVM is defined as:

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$$EVM = \frac{|V_{error}|}{|V_{reference}|}$$

where error vector is a vector between the ideal point and the real received point by the receiver in the constellation diagram.



Fig. 4. Constellation diagram of 2 streams demodulated from 802.11n signals received by MSAA

Fig. 5 and Fig. 6 show the CDF of the EVM of MSAA with the 0.2 and 0.6 wavelengths array spacing in LOS and NLOS environment, respectively. It is shown that the CDF of the EVM of stream 1 and stream 2 will be changed along with various array spacing. Moreover, the difference between the stream 1 and stream 2 at CDF=50% is also changed along with the different array spacing. When array spacing was increased, the difference will be almost decreased. And the EVM of the stream 1 is lower than that of the stream 2 because of the low channel gain of the MSE channel.





Fig. 5. CDF of EVM of MSAA with the 0.2 and 0.6 wavelengths array spacing in the LOS environment



Fig. 6. CDF of EVM of MSAA with the 0.2 and 0.6 wavelengths array spacing in the NLOS environment

Fig. 7 and Fig. 8 show the CDF of the EVM of the stream 1 and stream 2 along with various array spacing in LOS and NLOS environment, respectively. In the case of LOS and NLOS environment, it is found that when the array spacing is increased, both the EVM of the stream 1 and the stream 2 become large in the range of 0.1-0.6 wavelengths. The EVM of 2 streams with various array spacing are compared at CDF=50% in the LOS and the NLOS environment, respectively. It is shown that the EVM of 2 streams can be improved by decreasing the array spacing as small as 0.1 wavelengths in the case of LOS and NLOS environment.



Fig. 7. CDF of EVM of the stream 1 and stream 2 with various array spacing in LOS environment



Fig. 8. CDF of EVM of the stream 1 and stream 2 with various array spacing in NLOS environment

The channel capacity was also measured for evaluating MIMO performance. The channel capacity can be expressed as

$$C = \log_2 \left| I_{M_0} + \frac{P_{Total}}{M\sigma_n^2} \mathbf{H} \mathbf{H}^{\dagger} \right|$$
$$= \sum_{i=1}^{M_0} \log_2 \left(1 + \frac{P_{Total}}{M\sigma_n^2} \lambda_i \right) \quad M_0 = \min(M, N)$$

where \dagger for transpose conjugate, I_{M_0} for the $M_0 \times M_0$ identity matrix, P_{Total} is the total transmission power, σ_n^2 is the noise power, H is the MIMO channel matrix, λ_i is the ith eigenvalue, M is the number of the transmitting antennas and N is the number of the receiving antennas.

Fig. 9 shows the result of the MIMO channel capacity in the LOS environment. Similarly, the result of the NLOS environment was also shown as Fig. 10. It is found that when the array spacing is increased, the channel capacity become large.



Fig. 9. The MIMO channel capacity in the LOS environment



Fig. 10. The MIMO channel capacity in the NLOS environment

The experimental results indicate that the EVM and the channel capacity which reflects the MIMO performance can be affected by the array spacing in the range of 0.1-0.6 wavelengths in LOS and NLOS environment. The EVM result is improved by decreasing array spacing in the range of 0.1-0.6 wavelengths. Contrarily, the MIMO channel capacity is reduced by decreasing the array spacing at the same range. The EVM and the channel capacity with various array spacing are compared at CDF=50% in the LOS and NLOS environment, respectively. The result shows the EVM of the NLOS is better than that of the LOS. The other hand is that the channel capacity of the LOS is better than

that of the NLOS.

4. CONCLUSIONS

In this report, an experimental measurement has been carried out to study the MIMO performance in the LOS and NLOS environment of an indoor 2 by 2 MIMO system where the MSAA was used as the receiving antenna. The EVM and the channel capacity which reflects the MIMO performance were measured and compared along with the different array spacing to study the relation between the array spacing and the MIMO performance of the MSAA. The result has shown that EVM can be improved by decreasing the array spacing in the range of 0.1-0.6 in LOS and NLOS environment. Contrarily, the MIMO channel capacity will be reduced by decreasing the array spacing in the same range. So the EVM is complementary with the channel capacity.

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